

Comparison of PAPR Reduction Companding Techniques of OFDM Signal

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Abstract - the major drawbacks which is yet to be solved in orthogonal frequency division multiplexing (OFDM) system is its high peak-to-average power ratio (PAPR). In this paper, different companding transform scheme are compared which are nonlinear in nature and which transforms the Gaussian-distributed OFDM signals into a desired piecewise function. variable slopes and inflexion points is introduced in the target probability density function, to impart more flexibility in the companding form and an effective trade-off between the PAPR and bit error rate performances can be achieved along with a relatively low complexity. The theoretical analyses of different techniques and trade-off between PAPR reduction and bit error ratio (BER) performance are presented well.

Index Terms - orthogonal frequency division multiplexing (OFDM), nonlinear companding transform (NCT), peak-to-average power ratio (PAPR).

I. INTRODUCTION

Though orthogonal frequency division multiplexing (OFDM) offered significant advantages, the fluctuating envelope with high peaks is the major drawback of OFDM, which give rise to high peak to-average-power ratio (PAPR) in the transmitted signal. The transmitter's power amplifier (PA) drives into the saturation or nonlinear regions of operation because of High peaks , hence causing in-band distortions and out-of-band radiation. They also require wide dynamic ranges of analog-to-digital converters (ADC). Many techniques to reduce PAPR have been proposed in the literature, such as clipping and filtering, selective mapping ,companding transforms (CTs), partial transmit sequences, tone reservation, tone injection and linear block coding [1]–[3]. Complementary cumulative distribution function (ccdf) is used to measure PAPR reduction capability, which is defined as the probability that the signal's PAPR exceeds a specific threshold. In most methods, PAPR is reduced at the expense of increasing the bit error rate (BER), complexity, or data overhead. due to their flexibility and low complexity the CTs form an attractive and widely used PAPR reduction technique, with N subcarriers in an OFDM system, the complex baseband representation of OFDM signal is given by.

$$x(t) = \frac{1}{N} \sum_{k=0}^{N-1} X_k e^{j2\pi kt} \quad 0 \leq t \leq T \quad (1)$$

Irrespective of the number of OFDM signal subcarriers. CTs amplify the low amplitudes and attenuate the high peaks, so, by enlarging small signals and compressing large ones, both immunity of small signals from noise and the PAPR reduction can be achieved [6]. However, it is important to note that NCT is extra pre-distortion operation which is, applied to transmitted signal, which results in increased sensitivity to the HPA and performance degradation.

It was noticed in [8] that, due to the disadvantages of nonlinear distortion, such transform should be designed judiciously such that the clipped signal is as little as possible For this reason, how to reallocate the statistics of OFDM signal as well as the power more reasonably to reduce the companding distortion impact is the key challenge for a to design NCT method. Moreover, a flexible and effective trade-off among the overall performance of OFDM system with respect to PAPR reduction (power efficiency), spectral re-growth (bandwidth efficiency), bit error rate (BER) and also the implementation complexity should be considered.

In this paper, further motivated by the observation above the NCT algorithms are compared in which the Gaussian distributed signal is transformed into a desirable distribution form defined by piecewise functions with inflexion points. Compared to the methods discussed and the impact of companding distortion on the BER performance is studied. In addition the achievable PAPR reduction analytical expressions, the selection criteria of transform parameters and signal attenuation factor studied. The rest of this paper is outlined as follows. Section II briefly gives the idea about the statistical characteristic of OFDM signal, while in Section III the generic formulas for different algorithm are shown. performance study is given in Section IV section V shows comparison result, followed by the conclusions in Section V

Notation: The expectation and maximal element operator are denoted by $E \{.\}$ and $\max \{.\}$. We use $[.]^T$, $(.)^{-1}$ and $|.|$ to denote the transpose, inverse and modulus operation, respectively. $\text{sgn}(\cdot)$ stands for the sign function $\text{IFFT}_N\{.\}$ represents the N -point inverse fast Fourier transform (IFFT) operation. $\text{Prob}\{A\}$ is the probability of the event A.

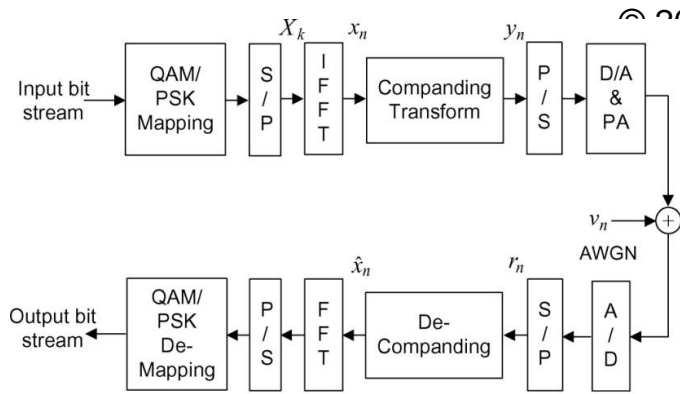


Fig. 1 Block diagram of OFDM system with companding transform

II. CHARACTERIZATION OF OFDM SIGNAL

Generally, an OFDM signal is the addition of N independent data symbols modulated by quadrature amplitude modulation (QAM) or phase-shift keying (PSK). In discrete-time domain, since the peaks of the continuous-time signal may not be represented by the Nyquist rate samples, so it is preferred to approximate on an oversampled signal. The time-domain oversampled OFDM symbols x can be calculated as

$$x = [x_0, x_1, \dots, x_{JN-1}]^T$$

$$x(n) = \frac{1}{N} \sum_{k=0}^{N-1} X_k e^{j2\pi kn} ; 0 \leq t \leq T \quad (2)$$

where n=0,1,2,...,JN-1 is time index and J is the oversampling ratio. Usually, J>=4 is used for accurate PAPR of the continuous-time signal. This process of oversampling can be achieved by performing a JN-point IFFT with extending X to a JN-length vector by inserting (j - 1) N extra zeros in its middle, i.e.

$$X_s = [X_0, \dots, X_{\frac{N}{2}-2}, 0, \dots, 0, X_{\frac{N}{2}}, \dots, X_{N-1}]^T \quad (3)$$

It is clear that $x = IFFT_{JN}\{X_s\}$. For a large N (i.e. N>=64), the real and imaginary parts of x_n may be approximated as Gaussian random variables with zero mean and a variance σ^2 . Based on this assumption, the signal amplitude $|x_n|$ follows a Rayleigh distribution with the probability density function (PDF) as

$$f_{|x_n|}(x) = \frac{2x}{\sigma^2} e^{-\frac{x^2}{\sigma^2}}, x \geq 0 \quad (4)$$

The cumulative density function (CDF) of $|x_n|$ is therefore

$$F_{|x_n|}(x) = prob\{|x_n| \leq x\} = \int_0^x \frac{2y}{\sigma^2} ep\left(-\frac{y^2}{\sigma^2}\right) dy = 1 - \exp\left(-\frac{x^2}{\sigma^2}\right), x \geq 0 \quad (5)$$

The PAPR of OFDM signal in a given frame is defined as

$$PAPR(x(t)) = \frac{\max [|x(t)|^2]}{P_{av}} \quad (6)$$

Considering the PAPR as a random variable is more helpful and utilizing a statistical description given by the complementary cumulative density function (CCDF), it is defined as the probability that the PAPR of exceeds an assigned level γ_0 . i.e

$$CCDF_x(\gamma_0) = Prob(PAPR_x > \gamma_0) \approx 1 - (1 - \exp(-\gamma_0))^N \quad (7)$$

The NCT principle is described as follows. The original signal x_n is companded before converted into analog waveform and amplified by the HPA. The signal after companding is denoted as $y_n = h(x_n)$, where $h(\cdot)$ is the companding function that only changes the amplitude of x_n . In additive Gaussian white noise (AWGN) channel case the received signal $r_n = y_n + v_n$ can be recovered using the de-companding function $h^{-1}(\cdot)$, namely, $x'_n = h^{-1}(y_n + v_n) = x_n + h^{-1}(v_n)$ where channel noise is v_n .

III. THE GENERIC FORMULAS FOR DIFFERENT ALGORITHM DESCRIPTION

Companding functions for different technique are as follows Pdf function for companding functions are represented by $f(\cdot)$ and companding function by $h(\cdot)$ [27], [28].

$$f_{|y_n|}(x) = \begin{cases} k_1 x, & 0 \leq x \leq cA \\ k_2 x + (k_1 - k_2) cA, & cA < x \leq A \end{cases} \quad (8)$$

$$h_1(x) = \begin{cases} \operatorname{sgn}(x) \sqrt{\frac{2}{k_1} \left(1 - \exp\left(-\frac{|x|^2}{\sigma^2}\right)\right)}, & |x| \leq X_0 \\ \operatorname{sgn}(x) \frac{1}{k_2} \left((k_2 - k_1)cA + \sqrt{(k_2 - k_1)k_1 c^2 A^2 + 2k_2 \left(1 - \exp\left(-\frac{|x|^2}{\sigma^2}\right)\right)} \right), & |x| > X_0 \end{cases} \quad (9)$$

this impact. Obviously, smaller value of companding distortion gives the reduced BER performance.

V. COMPARISON RESULTS

To compare the overall system performance of algorithms, on an OFDM system with N 1024 subcarriers 16 QAM modulation is used under AWGN channel.

$$f_{2|y_n}(x) = \begin{cases} \frac{2x}{\sigma^2} \exp\left(-\frac{x^2}{\sigma^2}\right), & 0 \leq x \leq c\sigma \\ kx + \frac{2c}{\sigma} \exp(-c^2 - kc\sigma), & c\sigma < x \leq A \end{cases} \quad (10)$$

$$h_2(x) = \begin{cases} x, & |x| \leq c\sigma \\ \operatorname{sgn}(x) \frac{1}{k} \left(kc\sigma - \frac{2c}{\sigma} \exp(-c^2) + \sqrt{\frac{4c^2}{\sigma^2} \exp(-2c^2) - \exp\left(-\frac{|x|^2}{\sigma^2}\right)} \right), & |x| > c\sigma \end{cases} \quad (11)$$

IV PERFORMANCE STUDY

In this section, the theoretical performance of algorithms are characterized with two main evaluation criteria the achievable reduction in PAPR and the impact of companding distortion on the BER performance at the receiver.

A. Achievable Reduction in PAPR

By making appropriate substitution in (6), the ultimate PAPR of the companded signal with the new algorithm is given by

$$PAPR_y = \frac{\max\left[|y_n|^2\right]}{E\left[|y_n|^2\right]} \quad (12)$$

Furthermore, a transform gain G is defined as the ratio of the PAPR of original signal to that of the companded signal [8], i.e.

$$G(db) = 10 \log_{10} \frac{PAPR_x}{PAPR_y} \quad (13)$$

These algorithms offer an adequate flexibility in the PAPR reduction by adjusting the values of k, k₂ and c. Consequently, the ultimate PAPR of the companded signal can be effectively confined in the interval [4.1 dB, 5.7 dB], or in other words, the achievable transform gain G in the PAPR is from 6 dB to 7.7 dB.

B. Impact of Companding Distortion

In essence, NCT is an extra nonlinear operation applied to the transmitted signal. For this reason, how to minimize the impact of companding distortion on the BER performance is the key in choosing the optimal companding form and parameters. Based on the analysis results for the Gaussian signals in [15] and [16], two performance criteria: signal attenuation and companding noise can be used to characterize

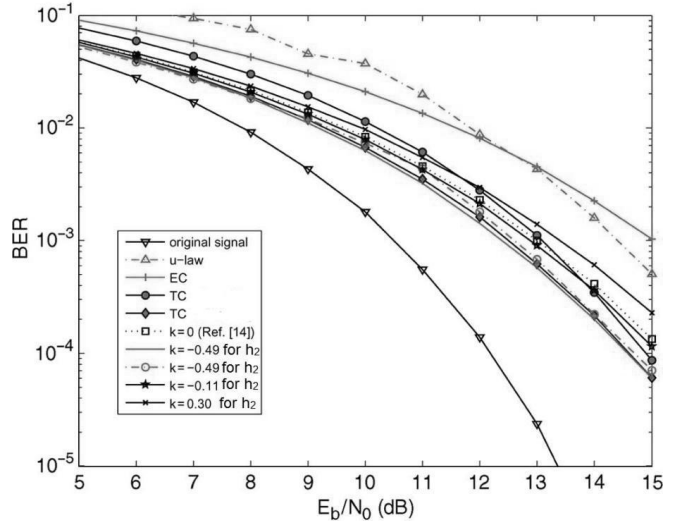


Fig. 2 Block BER performance of different transforms under an AWGN channel for N=1024 of the OFDM system with and 16 QPSK modulation. For h₁ companding function

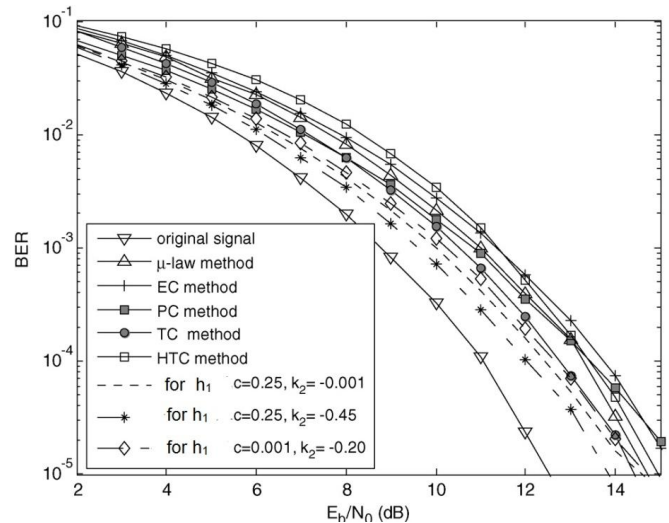


Fig. 3 BER performance of different transforms under an AWGN channel for N=1024 the OFDM system with and 16 QPSK modulation. For h₂ companding function.

For companding function h₁ and h₂ BER VS E_b/N₀ table is shown below

TABLE I

BER (dB)	$\frac{E_b}{N_0}$ for h_1 (C=0.25, $k_2=-0.45$)	$\frac{E_b}{N_0}$ for h_2 K=-0.45
11	10^{-2}	$10^{-1.6}$
12	10^{-3}	$10^{-2.8}$
13	$10^{-3.7}$	$10^{-3.3}$
14	$10^{-4.5}$	10^{-4}

VI CONCLUSIONS

Due to its simplicity and effectiveness, NCT is an attractive solution to reduce the PAPR of OFDM signal. In this paper, we compared NCT algorithm which changes the statistics of original signal from the complex Gaussian to a desirable PDF defined as a linear piecewise function. Thus, an effective and flexible trade-off between the PAPR and BER performance can be achieved to satisfy various system requirements. The discussed algorithm substantially outperforms the existing NCT methods in the overall performance of OFDM system regarding the reduction in PAPR, BER and out-of-band interference under the multipath fading channel or with the HPA.

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