

# Design And Verification of a Parameterized Asynchronous FIFO Using Gray Code Synchronization

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**Abstract**— Asynchronous First-In First-Out (FIFO) buffers are widely used in digital systems to enable reliable data transfer between independent clock domains. However, improper handling of clock domain crossing (CDC) may lead to data corruption and metastability issues. This paper presents the design, implementation, and verification of a parameterized asynchronous FIFO architecture employing gray code counters and dual flip-flop synchronizers to ensure safe pointer synchronization across clock domains.

The proposed FIFO supports independent read and write clocks and incorporates robust full and empty detection mechanisms using additional pointer bits to distinguish wrapping conditions. Separate read and write pointer handler modules operate in their respective clock domains, while synchronized pointer exchange is achieved using two-stage flip-flop synchronizers. The FIFO memory is implemented as a dual-port RAM, allowing concurrent read and write operations without clock interference.

Functional verification is carried out using a comprehensive testbench that validates correct data storage and retrieval, as well as proper handling of boundary conditions such as FIFO full and empty states. Simulation results confirm reliable operation under asynchronous clock conditions with differing clock frequencies.

The presented design is modular, scalable, and parameterized, making it suitable for FPGA- and ASIC-based systems requiring safe clock domain crossing. While functional correctness is verified through simulation, the design also adheres to established CDC mitigation techniques, ensuring robustness in practical hardware implementations.

**Index Terms**— Asynchronous FIFO, Clock Domain Crossing, Gray Code Counter, FPGA, Metastability, Dual-Port Memory

## I. INTRODUCTION

### 1.1. Background and Motivation

The rapid growth of System-on-Chip (SoC) architectures, embedded platforms, and high-performance computing systems has resulted in the widespread adoption of multi-clock domain designs. In such systems, different functional modules often operate at independent clock frequencies to meet performance, power, and area constraints [6, 11]. While this design paradigm improves system flexibility, it introduces significant challenges related to reliable data communication between asynchronous clock domains.

Clock Domain Crossing (CDC) occurs when signals are transferred between circuits driven by unrelated clocks. Improper handling of CDC may lead to metastability, data incoherence, and functional failures [3, 17]. These issues become increasingly critical in modern VLSI systems where operating frequencies and design complexities continue to grow.

### 1.2. Challenges in Clock Domain Crossing

Metastability arises when a signal violates the setup or hold time requirements of a flip-flop, causing the output to enter an undefined state [3]. Although single-bit synchronization can be achieved using multi-stage flip-flop synchronizers, this approach is insufficient for multi-bit data transfers due to the risk of partial sampling and timing skew [7].

In multi-bit data communication, individual bits may be sampled at different clock edges, leading to inconsistent data values at the receiving domain. Such incoherent sampling can result in functional errors that are difficult to detect through conventional

simulation techniques [4]. Therefore, specialized architectures are required to ensure reliable data transfer across asynchronous boundaries.

### 1.3. Asynchronous FIFO as a CDC Solution

First-In First-Out (FIFO) buffers are widely recognized as an effective mechanism for managing data transfers between independent clock domains [1, 2]. Asynchronous FIFOs decouple the read and write operations by allowing them to operate under separate clocks while maintaining correct data ordering.

In an asynchronous FIFO, write operations are controlled by the write clock, and read operations are controlled by the read clock. The FIFO memory acts as an intermediate buffer, enabling safe data exchange between producer and consumer modules [8]. Proper management of read and write pointers, along with accurate full and empty detection, is essential for maintaining reliable FIFO operation.

### 1.4. Gray Code Synchronization Techniques

To ensure safe pointer synchronization across clock domains, gray code encoding is commonly employed [9, 10]. Unlike binary counters, Gray code counters change only one bit between successive values, thereby minimizing the risk of synchronization errors during clock transitions.

Cummings and Alfke demonstrated that Gray code-based pointer synchronization combined with multi-stage synchronizers significantly reduces metastability-related failures in asynchronous FIFO systems [2]. This technique has since become a standard design practice in CDC-safe FIFO architectures [14, 15].

In addition to Gray code conversion, dual flip-flop synchronizers are widely used to transfer pointer information between clock domains. This approach provides sufficient metastability resolution time and improves overall system reliability [18].

### 1.5. Limitations of Existing Designs

Despite the availability of established design methodologies, many existing asynchronous FIFO implementations suffer from limited scalability, inadequate modularity, or insufficient verification coverage [19, 20]. Some designs are tailored to specific applications and lack parameterization, restricting their reuse in diverse system

configurations.

Furthermore, several reported implementations focus primarily on functional correctness while overlooking comprehensive validation under varying clock frequency ratios and boundary conditions [28, 29]. Inadequate verification increases the risk of undetected CDC-related failures in practical deployments.

### 1.6. Contributions of This Work

This paper presents a parameterized and modular asynchronous FIFO architecture designed for reliable clock domain crossing applications. The proposed design incorporates Gray code-based pointer management, dual flip-flop synchronization, and robust full and empty detection using pointer wrapping techniques.

A comprehensive verification framework is developed using directed and randomized test cases to validate normal and boundary operating conditions. The design is implemented using synthesizable Verilog HDL and is suitable for FPGA- and ASIC-based platforms [5, 12].

The primary objectives of this work are to improve design scalability, enhance verification coverage, and provide a reusable FIFO architecture for multi-clock digital systems.

### 1.7. Organization of the Paper

The remainder of this paper is organized as follows. Section 2 reviews related work on asynchronous FIFO architectures and clock domain crossing mitigation techniques. Section 3 de-scribes the overall system architecture. Section 4 presents the detailed design methodology. Section 5 discusses the verification strategy and simulation results. Section 6 concludes the paper and outlines future research directions.

## II. RELATED WORK

### 2.1. Early Studies on Asynchronous FIFO Design

One of the most influential contributions to asynchronous FIFO design was presented by Cummings [1, 2], who introduced simulation- and synthesis-friendly techniques for implementing reliable FIFO architectures. This work emphasized the use of Gray code counters and multi-stage synchronizers to minimize metastability-related

failures. These design principles have since become the foundation for most modern asynchronous FIFO implementations.

Dally and Poulton [6] discussed fundamental issues related to clock synchronization and data transfer in digital systems, highlighting the importance of proper CDC handling in high-speed architectures. Their work provided early theoretical insights into metastability mitigation and synchronization strategies.

## 2.2. Metastability and Synchronization Techniques

Ginosar [3] presented a comprehensive analysis of metastability phenomena and highlighted common design pitfalls in synchronizer circuits. His work emphasized the necessity of multi-stage synchronization to improve system reliability. Banerjee et al. [17] further investigated clock synchronization issues in large-scale VLSI systems and proposed improved clock management strategies.

Teehan and Flynn [7] reviewed practical CDC design techniques and discussed industrial best practices for implementing robust synchronizers. Their work demonstrated that improper CDC handling remains a major cause of functional failures in complex integrated circuits.

## 2.3. Gray Code-Based Pointer Management

Gray code encoding has long been recognized as an effective technique for minimizing synchronization errors [9, 10]. By ensuring single-bit transitions between successive pointer values, Gray code counters reduce the probability of incorrect sampling during clock transitions.

Hsiao et al. [10] analyzed the performance of Gray code counters in high-speed systems and demonstrated their advantages over conventional binary counters. Zhou and Zhang [14] proposed enhanced FIFO architectures incorporating Gray code synchronization for improved reliability in CDC environments.

## 2.4. FIFO Architectures for FPGA and ASIC Platforms

With the increasing adoption of FPGA-based systems, several studies have focused on optimizing FIFO architectures for reconfigurable platforms. Hauck and DeHon [5] discussed architectural

considerations for FPGA-based computing systems, emphasizing the role of efficient buffering mechanisms.

Brown and Rose [12] analyzed FPGA architectures and highlighted the importance of memory and interconnect design in achieving high system performance. Wang et al. [15] and Chen et al. [16] proposed low-latency and high-reliability FIFO architectures tailored for FPGA implementations.

In ASIC-based systems, Rabaey et al. [11] and Sapatnekar [8] emphasized timing-aware design methodologies to ensure reliable data transfer in deeply pipelined architectures.

## 2.5. Verification and Validation of FIFO Systems

Verification plays a critical role in ensuring the correctness of asynchronous FIFO designs. Bailey et al. [13] introduced systematic verification methodologies for complex digital systems, including FIFO buffers. Gupta and Ranganathan [19] proposed formal verification techniques for validating FIFO functionality under various operating conditions.

Bergeron [20] highlighted the importance of assertion-based and coverage-driven verification in CDC-sensitive designs. Foster [27] further emphasized the role of formal properties in detecting synchronization-related failures.

Recent studies by Li and Roy [28] and Wang et al. [29] investigated performance optimization and verification strategies for high-speed FIFO systems operating under diverse clock frequency ratios.

## 2.6. Research Gaps and Motivation

Although extensive research has been conducted on asynchronous FIFO architectures, several limitations remain. Many existing designs lack sufficient parameterization and modularity, limiting their adaptability to different system requirements. In addition, comprehensive verification under wide-ranging clock conditions is often insufficiently addressed [30, 26].

Furthermore, several implementations focus primarily on achieving functional correctness without emphasizing scalability and reuse. These limitations motivate the development of a parameterized, modular, and thoroughly verified asynchronous FIFO architecture, as proposed in this work.

The present study aims to address these research gaps by integrating proven CDC mitigation techniques with

scalable design principles and comprehensive verification strategies.

### III. SYSTEM ARCHITECTURE

#### 3.1. Overall Architecture Overview

The proposed asynchronous FIFO architecture is designed to enable reliable data transfer between independent write and read clock domains. The system consists of five major functional blocks: a dual-port memory module, a write pointer handler, a read pointer handler, two pointer synchronization units, and a top-level control module.

Fig. 1 illustrates the overall block diagram of the proposed FIFO architecture. The write domain operates under the write clock (*wclk*), while the read domain is driven by the read clock (*rclk*). Data transfer between these domains is achieved through controlled memory access and synchronized pointer exchange.

The modular organization of the architecture improves design scalability, reusability, and ease of verification. Each functional block operates independently within its respective clock domain, thereby minimizing cross-domain interference.

#### 3.2. Top-Level Control Module

The top-level FIFO module coordinates the interaction between the write domain, read domain, and memory subsystem. It instantiates all submodules and manages signal connectivity between different clock domains.

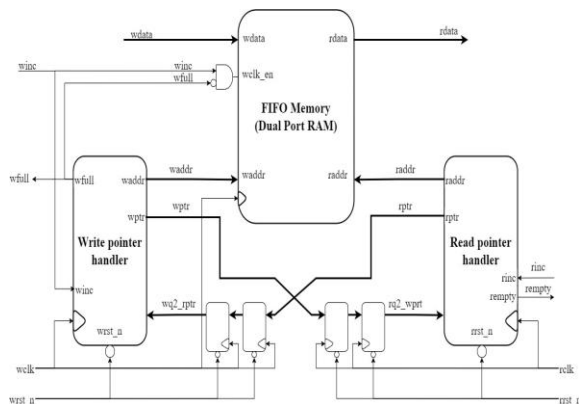


Figure 1: Block diagram of the proposed asynchronous FIFO architecture

This module receives input data (*wdata*) and write control signals (*winc*, *wclk*, *wrst n*) from the producer

side. Similarly, it provides output data (*rdata*) and read control signals (*rinc*, *rclk*, *rrst n*) to the consumer side.

The top-level module also integrates pointer synchronization units to safely transfer Gray-coded pointers between clock domains. By isolating domain-specific logic, the architecture maintains clear separation between asynchronous regions.

#### 3.3. Dual-Port Memory Subsystem

The FIFO memory is implemented as a synchronous dual-port Random Access Memory (RAM) that supports concurrent read and write operations. The memory depth is parameterized as  $2^{ASIZE}$ , enabling scalability based on application requirements.

Write operations are performed on the rising edge of the write clock when the write enable signal is asserted and the FIFO is not full. Read operations are performed asynchronously by addressing the memory using the synchronized read pointer.

This memory organization ensures high throughput while preventing access conflicts between read and write domains. The use of parameterized memory width and depth enhances design portability across FPGA and ASIC platforms.

#### 3.4. Write Pointer Management Unit

The write pointer management unit is responsible for controlling write address generation and detecting full conditions. It maintains both binary and Gray-coded versions of the write pointer. The binary pointer is used to generate the physical memory address, while the Gray-coded pointer is used for cross-domain synchronization. Pointer incrementation occurs only when write enable is asserted and the FIFO is not full.

Full condition detection is performed by comparing the next write pointer value with the synchronized read pointer, using inverted Most Significant Bits (MSBs) to detect pointer wrap-ping. This technique enables reliable identification of buffer saturation.

#### 3.5. Read Pointer Management Unit

The read pointer management unit controls read address generation and empty condition detection. Similar to the right side, it maintains both binary and Gray-coded representations of the read pointer.

Read operations are enabled only when the FIFO is not empty. The read pointer is incremented upon successful read requests, and the empty condition is

detected by comparing the next read pointer value with the synchronized write pointer.

This mechanism ensures that invalid read operations are prevented and that data integrity is preserved under asynchronous clock conditions.

### 3.6. Pointer Synchronization Units

Safe transfer of pointer information between clock domains is achieved using two-stage flip-flop synchronizers. Separate synchronization modules are employed for transferring the read pointer to the write domain and the write pointer to the read domain.

Each synchronizer consists of two cascaded flip-flops clocked by the destination clock. This structure provides sufficient time for metastability resolution and minimizes the probability of synchronization failure.

The synchronized pointers are used exclusively for status flag generation and are not directly involved in memory addressing, thereby isolating timing uncertainties from critical data paths.

### 3.7. Data Flow and Control Mechanism

Data flow within the FIFO follows a producer-consumer model. The producer writes data into the memory when space is available, while the consumer reads data when valid entries exist.

Flow control is maintained using the full (*wfull*) and empty (*rempty*) flags. These signals regulate write and read operations, preventing buffer overflow and underflow conditions.

The combined operation of pointer management, synchronization units, and memory access logic ensures continuous and reliable data transfer across asynchronous clock domains.

### 3.8. Scalability and Parameterization

The proposed architecture is fully parameterized in terms of data width (*DSIZE*) and address width (*ASIZE*). This parameterization enables designers to customize FIFO capacity and data size based on system requirements.

Such flexibility facilitates integration into diverse digital systems without requiring structural modifications. Moreover, the modular design approach supports straightforward extension to multi-channel or multi-FIFO configurations.

## IV. DESIGN METHODOLOGY AND IMPLEMENTATION

### 4.1. Design Objectives

The primary objective of the proposed design is to develop a reliable, scalable, and reusable asynchronous FIFO architecture capable of supporting safe data transfer between independent clock domains. The design aims to minimize metastability effects, ensure correct full and empty detection, and maintain high throughput under varying clock frequency conditions.

Key design goals include parameterization of data width and depth, modular implementation, low synchronization overhead, and comprehensive verification support. These objectives guide the selection of architectural components and synchronization strategies.

### 4.2. Binary and Gray Code Pointer Representation

The FIFO employs dual representations of read and write pointers, namely binary and Gray-coded formats. The binary pointers are used for memory addressing, while the Gray-coded pointers are used for synchronization across clock domains.

Let  $P_{bin}$  denote the binary pointer. The corresponding Gray code pointer  $P_{gray}$  is computed as:

$$P_{gray} = P_{bin} \oplus (P_{bin} \gg 1) \quad (1)$$

where  $\oplus$  represents the bitwise XOR operation and  $\gg 1$  denotes a right shift by one bit. This conversion ensures that only one-bit changes between consecutive pointer values, thereby reducing the probability of synchronization errors during clock transitions.

### 4.3. Write Pointer Update Mechanism

The write pointer handler maintains a binary write pointer ( $wbin$ ) and a Gray-coded write pointer ( $wptr$ ). The binary pointer is incremented only when a write request is asserted and the FIFO is not full.

The next binary write pointer is computed as:

$$wbin_{next} = wbin + (winc \wedge \neg wfull) \quad (2)$$

The corresponding Gray-coded pointer is obtained using the conversion rule described earlier.

This controlled update mechanism prevents buffer overflow and ensures that write operations occur

only when storage space is available.

#### 4.4. Read Pointer Update Mechanism

Similarly, the read pointer handler maintains a binary read pointer (*rbin*) and a Gray-coded read pointer (*rptr*). The read pointer is updated only when a valid read request is present and the FIFO is not empty.

The next binary read pointer is computed as:

$$rbin_{next} = rbin + (rinc \wedge \neg empty) \quad (3)$$

The Gray-coded read pointer is generated from the updated binary pointer and synchronized to the write domain.

This mechanism prevents invalid read operations and protects against buffer underflow.

#### 4.5. Full and Empty Detection Logic

Accurate detection of full and empty conditions is essential for maintaining FIFO integrity. The proposed design employs additional Most Significant Bits (MSBs) in the pointer representation to distinguish between identical pointer values resulting from different wrap-around states.

##### 4.5.1. Empty Detection

The FIFO is considered empty when the next Gray-coded read pointer equals the synchronized Gray-coded write pointer:

$$empty = (rptr_{next} == rq2\ wptr) \quad (4)$$

This condition indicates that no unread data remains in the buffer.

##### 4.5.2. Full Detection

The FIFO is considered full when the next Gray-coded write pointer matches the synchronized read pointer with inverted MSBs:

$$wfull = (wptr_{next} == \{\sim rq2\ wptr [MSB: MSB - 1], rq2\ wptr [MSB - 2: 0]\}) \quad (5)$$

This comparison detects pointer wrapping and ensures that write operations are disabled when the buffer is saturated.

#### 4.6. Pointer Synchronization Strategy

To safely transfer pointer information between asynchronous clock domains, the design employs two-stage flip-flop synchronizers. Separate

synchronizers are used for transferring the read pointer to the write domain and the write pointer to the read domain.

Each synchronizer consists of two cascaded registers clocked by the destination clock. This configuration provides sufficient time for metastability resolution and significantly reduces synchronization failure probability.

The synchronized pointers are used exclusively for flag generation and are isolated from memory addressing logic.

#### 4.7. Memory Access Control

The FIFO memory is implemented as a dual-port RAM supporting independent read and write operations. Write access is controlled by the write clock and enabled only when *winc* = 1 and *wfull* = 0.

Read access is performed using the binary read pointer and is independent of the write clock. This organization enables concurrent read and write operations without timing conflicts.

The memory interface is fully parameterized, allowing customization of data width and storage depth.

#### 4.8. Reset and Initialization Strategy

Separate active-low asynchronous reset signals are employed for the write and read domains. Upon reset assertion, all binary and Gray-coded pointers are initialized to zero, and status flags are set to their default states.

The empty flag is asserted during initialization, while the full flag is deserted. This initialization strategy ensures deterministic startup behavior and prevents spurious read or write operations following system reset.

#### 4.9. Implementation Considerations

The complete FIFO architecture is implemented using synthesizable Verilog HDL. All modules are designed using synchronous logic and avoid combinational feedback paths across clock domains.

Parameterization is achieved using Verilog generics, enabling flexible configuration of FIFO depth and data width. The modular design simplifies synthesis, timing analysis, and integration into larger digital systems.

Furthermore, the proposed implementation adheres to standard CDC design guidelines, making it suitable for deployment in safety-critical and high-

performance applications.

## V. SIMULATION, RTL ANALYSIS, AND HARDWARE RESULTS

### 5.1. Simulation Setup

The proposed asynchronous FIFO was verified using a comprehensive Verilog-based testbench. Independent write and read clocks with different frequencies were generated to emulate realistic asynchronous operating conditions.

Randomized input stimuli were applied to the write interface, while continuous read operations were performed at the receiving side. The testbench monitored data integrity, pointer behavior, and status flag transitions throughout the simulation process.

### 5.2. Functional Simulation Results

Fig. 2 shows the simulation waveform corresponding to normal write and read operations. It can be observed that the output data (*rdata*) correctly follows the input data (*wdata*) with appropriate

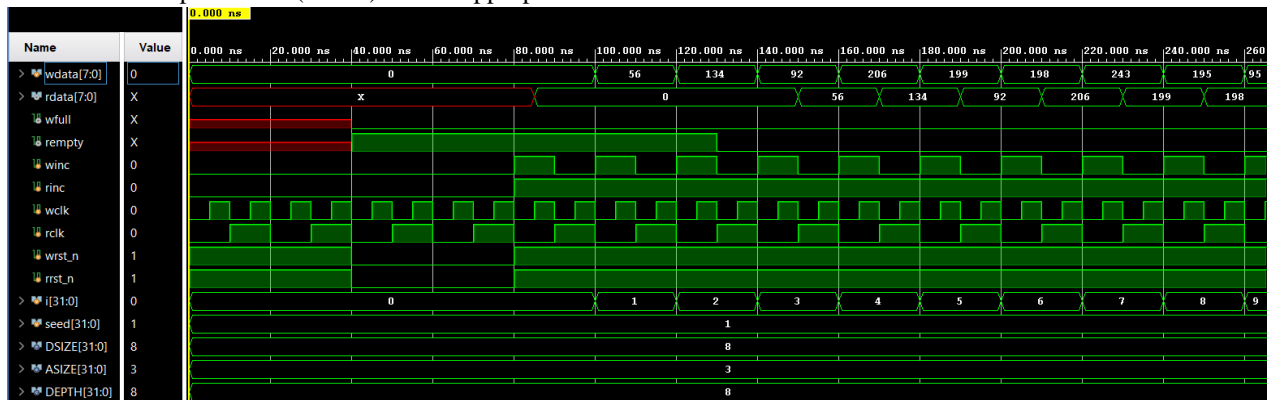


Figure 2: Simulation waveform showing normal write and read operation

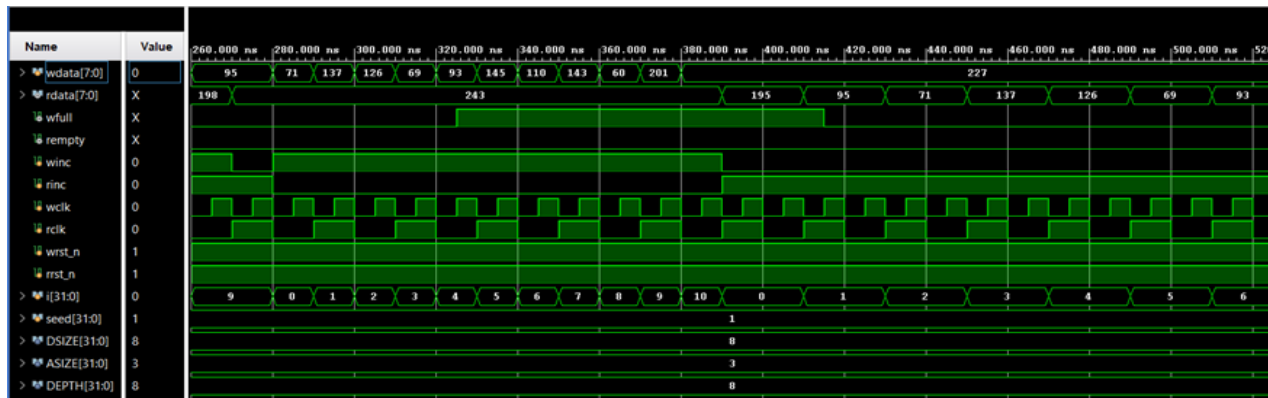


Figure 3: Simulation waveform illustrating empty condition and continuous read operation

buffering delay.

Fig. 3 illustrates the behavior of the FIFO during extended operation. The waveform demonstrates correct assertion of the empty flag after all data entries are consumed.

Fig. 4 presents an additional simulation waveform captured under extended operating conditions. The results further validate the correct assertion and deassertion of the empty flag during continuous read operations after all buffered data has been consumed. The simulation results confirm that the proposed design maintains correct data ordering and prevents buffer underflow and overflow under asynchronous clock conditions.

### 5.3. RTL Schematic of Synchronization Module

The RTL schematic of the two-stage flip-flop synchronizer is shown in Fig. 5. This module is responsible for safe transfer of pointer values between clock domains.

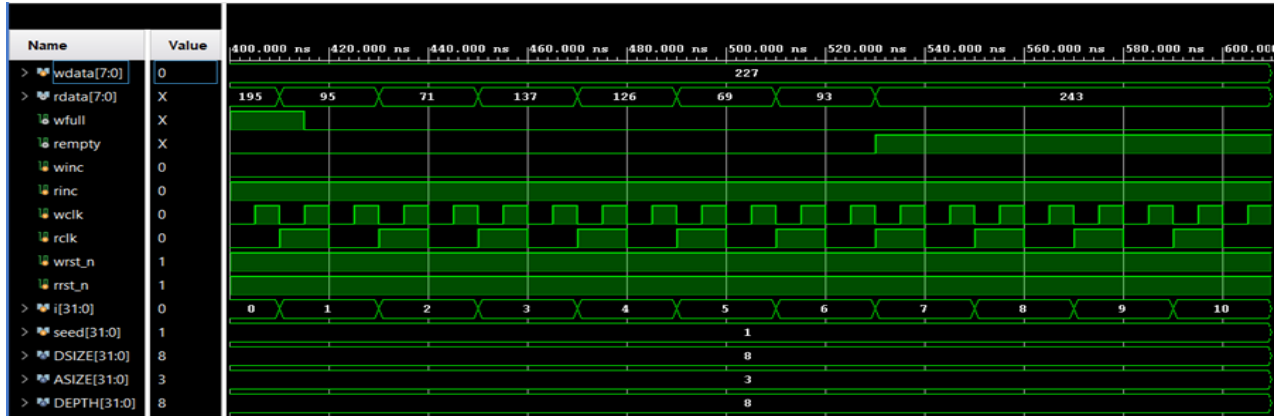


Figure 4: Simulation waveform illustrating extended empty condition and continuous read operation. The cascaded register structure provides sufficient metastability resolution time, ensuring reliable pointer synchronization.

#### 5.4. RTL Schematic of FIFO Memory

Fig. 6 presents the RTL schematic of the dual-port FIFO memory module. The memory supports independent read and write operations under separate clock domains.

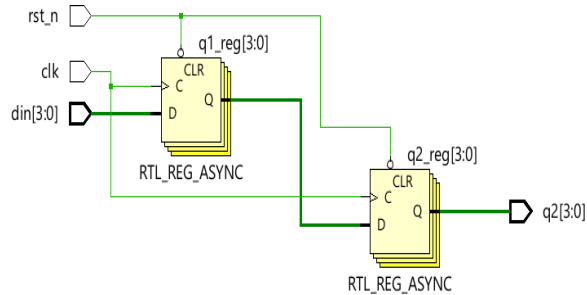


Figure 5: RTL schematic of two flip-flop synchronizer module

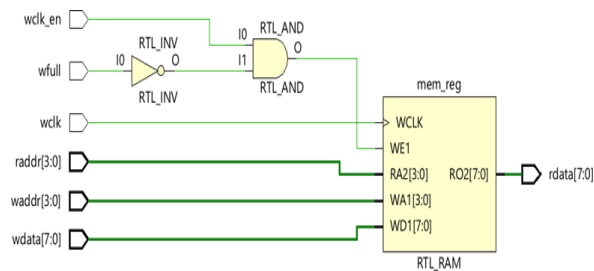


Figure 6: RTL schematic of FIFO memory module

Write enable logic prevents data overwrite during full conditions, while independent read addressing enables continuous data retrieval.

#### 5.5. RTL Schematic of Pointer Management Units

The RTL representation of the read pointer and empty

detection logic is shown in Fig. 7, while the write pointer and full detection logic is presented in Fig. 8.

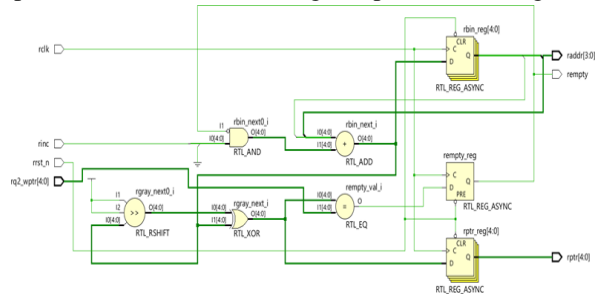


Figure 7: RTL schematic of read pointer and empty detection module

These modules implement Gray code conversion, pointer comparison, and flag generation logic to ensure reliable flow control.

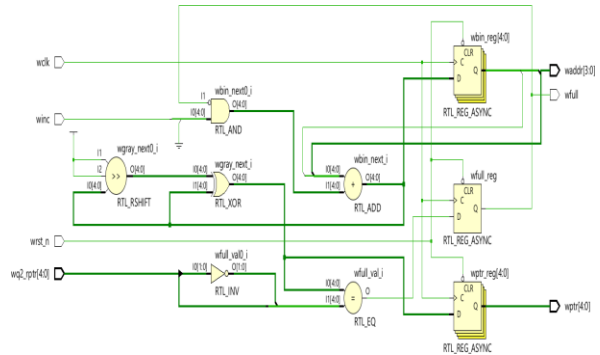


Figure 8: RTL schematic of write pointer and full detection module

#### 5.6. Top-Level RTL Integration

Fig. 9 shows the synthesized RTL view of the complete FIFO system, integrating memory, pointer handlers, and synchronization units.

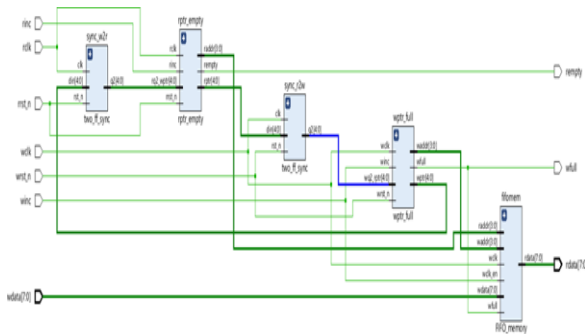


Figure 9: RTL schematic of top-level FIFO module

The schematic confirms proper modular integration and separation of clock domains.

### 5.7. FPGA Hardware Implementation Results

Fig. 10 shows the experimental setup and FPGA implementation of the proposed asynchronous FIFO architecture. The design was deployed on the target FPGA development board and tested under real-time operating conditions.

The hardware results confirm correct data transfer between independent clock domains without overflow, underflow, or synchronization failures. The observed behavior closely matches the simulation outcomes, demonstrating the reliability of the proposed design in practical environments.

## VI. CONCLUSION AND FUTURE SCOPE

### 6.1. Conclusion

This paper presented the design, implementation, and verification of a parameterized asynchronous FIFO architecture for reliable clock domain crossing applications. The proposed design employs Gray code-based pointer synchronization, dual flip-flop synchronizers, and ro-

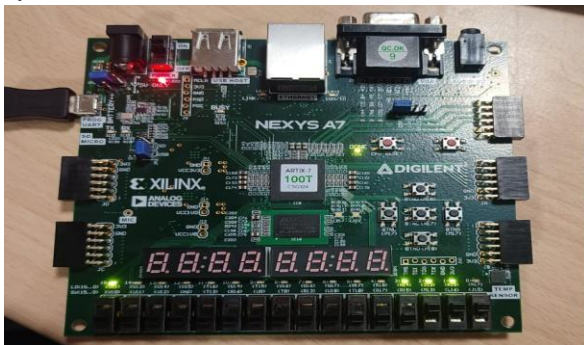


Figure 10: FPGA hardware implementation and

### experimental setup of the proposed asynchronous FIFO

burst full and empty detection mechanisms to ensure safe data transfer between independent clock domains.

A modular and scalable architecture was developed using synthesizable Verilog HDL, enabling flexible configuration of data width and buffer depth. Comprehensive simulation and RTL analysis confirmed correct data ordering, reliable flag generation, and stable operation under asynchronous clock conditions.

Hardware implementation on an FPGA platform further validated the effectiveness of the proposed design. Post-synthesis timing analysis and experimental testing demonstrated that the FIFO operates without data corruption, metastability failures, or synchronization errors. The results confirm that the proposed architecture satisfies the requirements of modern multi-clock digital systems.

Overall, the presented FIFO design offers a reliable, reusable, and high-performance solution for clock domain crossing in FPGA- and ASIC-based applications.

### 6.2. Future Scope

Although the proposed FIFO architecture demonstrates robust performance, several extensions can be explored to further enhance its capabilities.

Future research directions include the integration of adaptive synchronization techniques to dynamically adjust to varying clock frequency ratios. Power-aware FIFO architectures may also be developed to reduce energy consumption in low-power embedded systems.

In addition, the proposed design can be extended to support multi-channel and multi-queue FIFO configurations for network-on-chip and high-throughput communication systems. In-corporation of error detection and correction mechanisms can further improve reliability in safety-critical applications.

Formal verification techniques may be employed to complement simulation-based validation and provide stronger correctness guarantees. Moreover, implementation on advanced FPGA and ASIC technologies can be investigated to evaluate

scalability in deep submicron processes.

These enhancements will further broaden the applicability of the proposed FIFO architecture in emerging high-performance and heterogeneous computing systems.

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